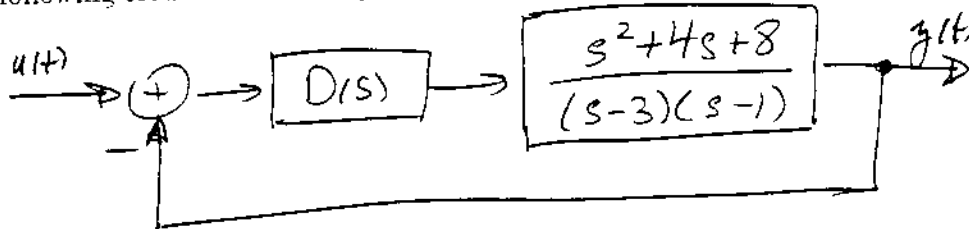


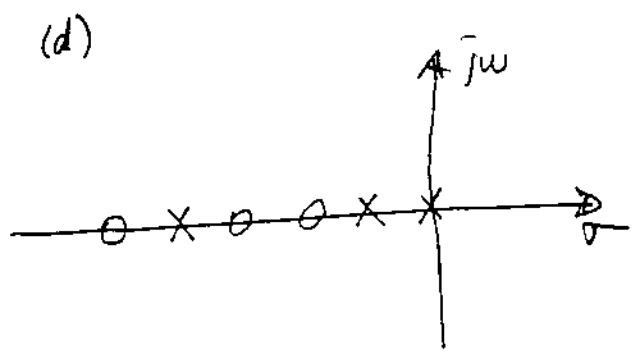
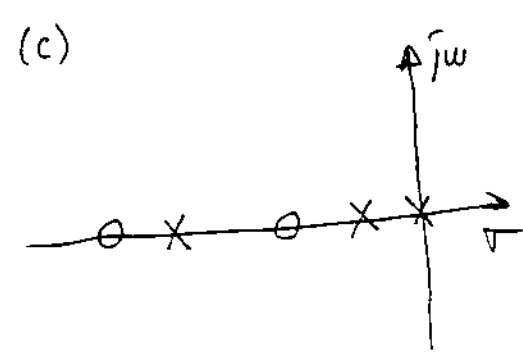
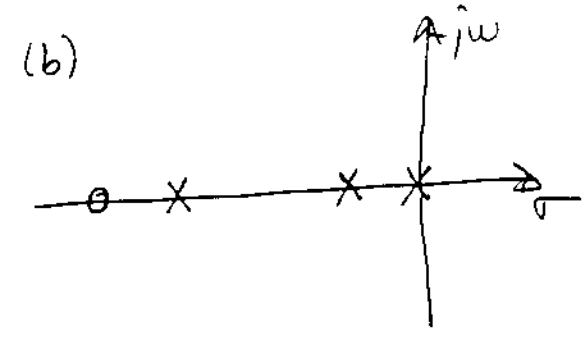
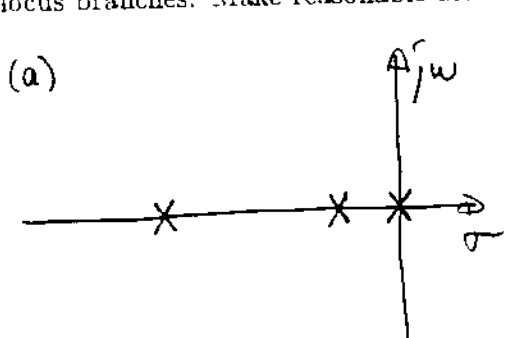
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1. Consider the following feedback control systems.



Note: This problem might require hand construction of some diagrams. In this case, the inaccuracies in the closed-loop pole locations are acceptable.

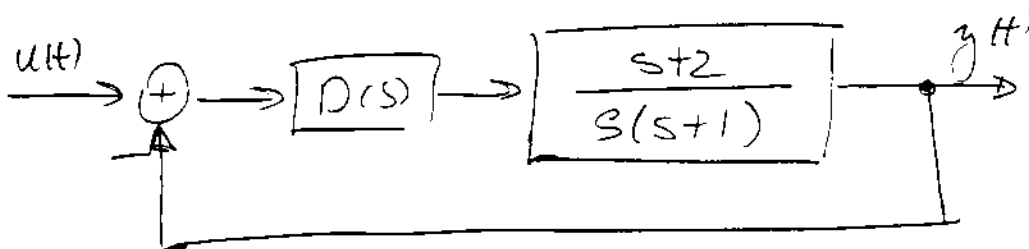
- (a) Design a proportional (P) controller, such that the normalized maximum overshoot is  $M_p \approx 0.25$ . (10pts)
  - (b) Design an integral (I) controller, such that the steady-state error for the unit ramp input is at most 0.1. (5pts)
  - (c) Design the simplest controller, such that the 2% settling time is at most 4 seconds. After the design, determine the normalized maximum overshoot. (10pts)
2. For the following open-loop pole/zero locations, sketch *expected* root-locus diagrams. Do not determine any features of the diagram, except the asymptote angles. Simply show the expected shapes of all the root-locus branches. Make reasonable assumptions for all the undetermined features. (20pts)



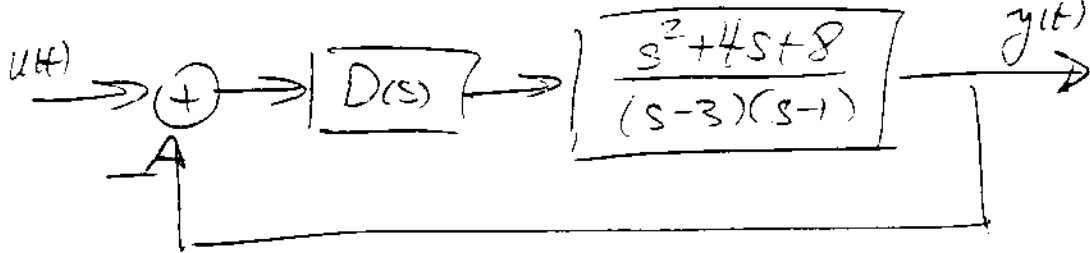
3. Consider a unity-feedback control system with the open-loop transfer function

$$G(s) = K \frac{1}{s(s+2)(s^2+6s+13)} = K \frac{\cancel{s^2+8s+17}}{s^4+8s^3+25s^2+26s}$$

- (a) Construct the root-locus diagram. Determine all the important features like asymptotes, imaginary-axis crossings, angle of arrivals and departures; however *do not* determine the break-away and/or break-in points explicitly. In other words, obtain the equation whose solutions would give those points, but *do not* solve that equation. (30pts)
- (b) Determine all the values of  $K$  such that the closed-loop system is asymptotically stable. (5pts)
4. For the following system, design a first order compensator  $D(s)$ , without increasing the order of the system, such that the steady-state error is zero for a step input, and the 5% settling time is approximately 0.5 second. (20pts)



#1

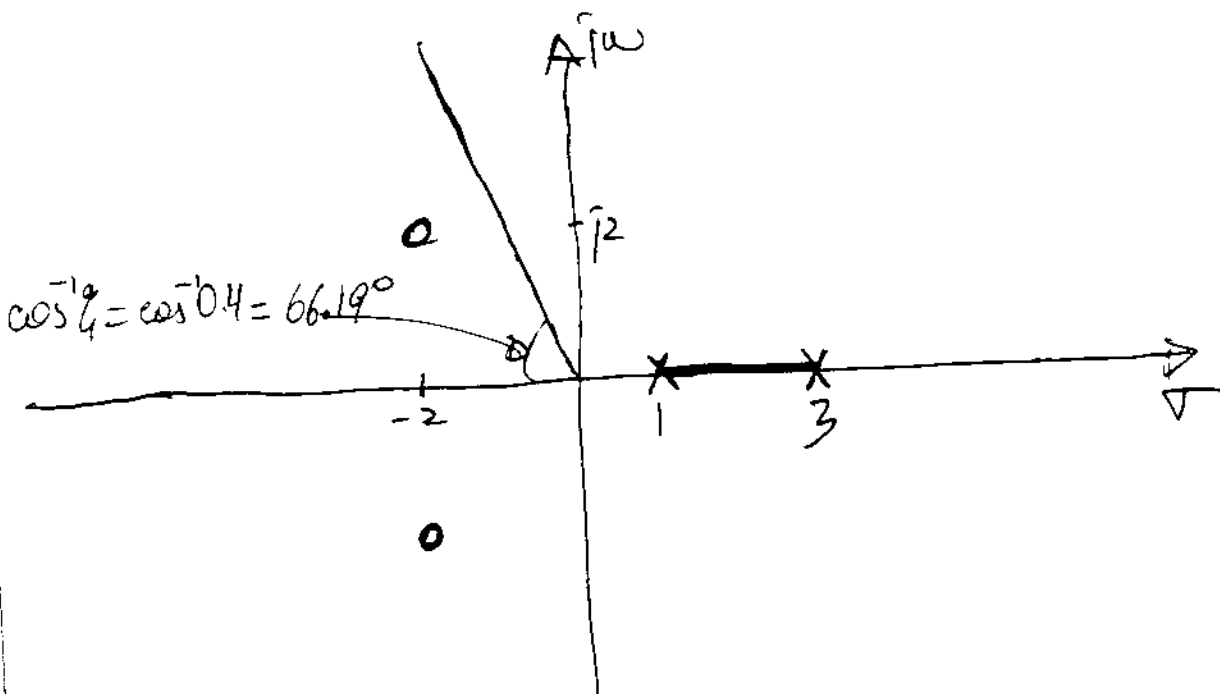


1// Proportional Control  $\Rightarrow D(s) = K$

$$GH = K \frac{s^2 + 4s + 8}{(s-3)(s-1)} \leftarrow s = -2 \pm j2$$

$$M_p \approx 0.25 \Rightarrow e^{-\frac{\zeta}{\sqrt{1-\zeta^2}} \pi} \approx 0.25$$

$$\zeta \approx \frac{|\ln 0.25|}{\sqrt{\ln^2 0.25 + \pi^2}} = 0.4$$



Need to find out the imaginary and crossing and the angle of arrival and the break away point

(For the purposes of the exam, it was O.K. to sketch the root-locus without finding the above pts.)

Imaginary axis crossing  $\rightarrow$  Routh table

$$1 + G H = 0$$

$$1 + K \frac{s^2 + 4s + 8}{(s-3)(s-1)} = 0$$

$$(s^2 - 4s + 3) + K(s^2 + 4s + 8) = 0$$

$$(K+1)s^2 + (4K-4)s + (8K+3) = 0$$

|       |        |                   |                   |                         |
|-------|--------|-------------------|-------------------|-------------------------|
| $s^2$ | $K+1$  | $8K+3$            | $\longrightarrow$ | $(K+1)s^2 + (8K+3) = 0$ |
| $s$   | $4K-4$ | $\longrightarrow$ | $4K-4=0, K=1$     | }                       |
| $1$   | $8K+3$ | $\downarrow$      |                   |                         |

$$2s^2 + 11 = 0$$

$$s = \pm j 2.3452$$

Angle of arrival  $\rightarrow$  Angular condition at  $s_0 = -2 + j2$

$$-\angle(s_0 - 1) - \angle(s_0 - 3) + \angle(s_0 + 2 + j2) + \theta_{in} = 180^\circ + k360^\circ$$

$$-\tan^{-1} \frac{2-0}{-2-1} - \tan^{-1} \frac{2-0}{-2-3} + 90^\circ + \theta_{in} = 180^\circ + k360^\circ$$

$$-146.31^\circ - 158.20^\circ + 90^\circ + \theta_{in} = 180^\circ + k360^\circ$$

$$\theta_{in} = 34.51^\circ$$

Break-away point  $\rightarrow \frac{dK}{ds} = 0$

$$1 + GH = 0$$

$$1 + K \frac{s^2 + 4s + 8}{(s-3)(s-1)} = 0 \quad -K = \frac{s^2 - 4s + 3}{s^2 + 4s + 8}$$

$$\frac{dK}{ds} = \frac{(2s-4)(s^2+4s+8) - (s^2-4s+3)(2s+4)}{(s^2+4s+8)^2}$$

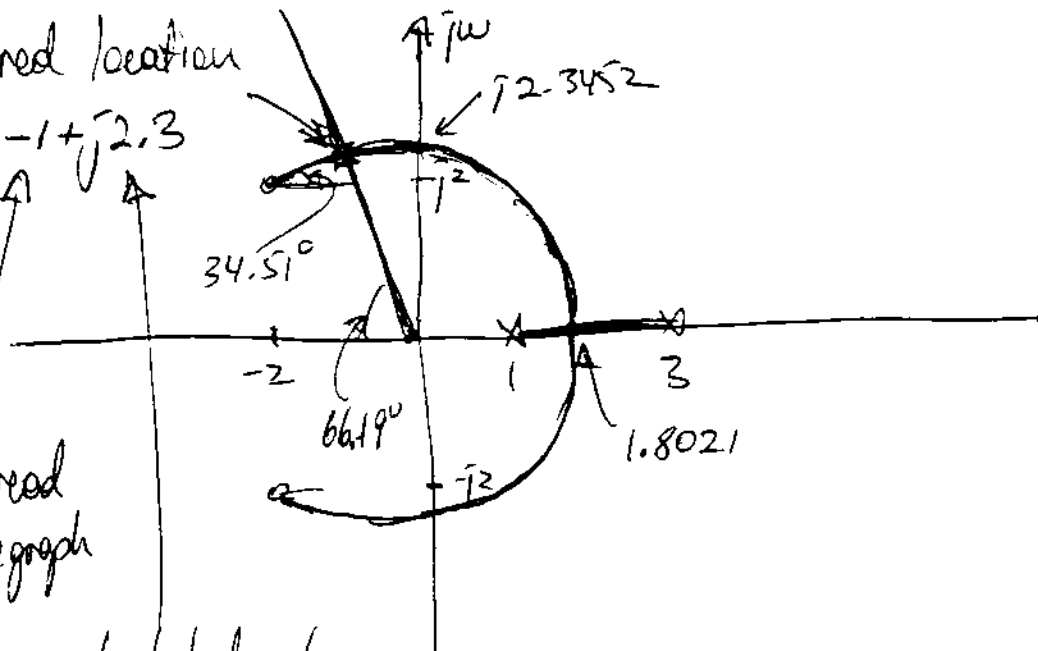
$$\frac{dK}{ds} = 0, \quad 8s^2 + 10s - 44 = 0$$

$$s = \frac{-5 \pm \sqrt{377}}{8} = 1.8021, -3.0521$$

Desired location

$$s_d \approx -1 + j2.3$$

approx. read  
from the graph



calculated s.t.  
sd has  $\zeta = 0.4$   
from  $\omega_n$ .

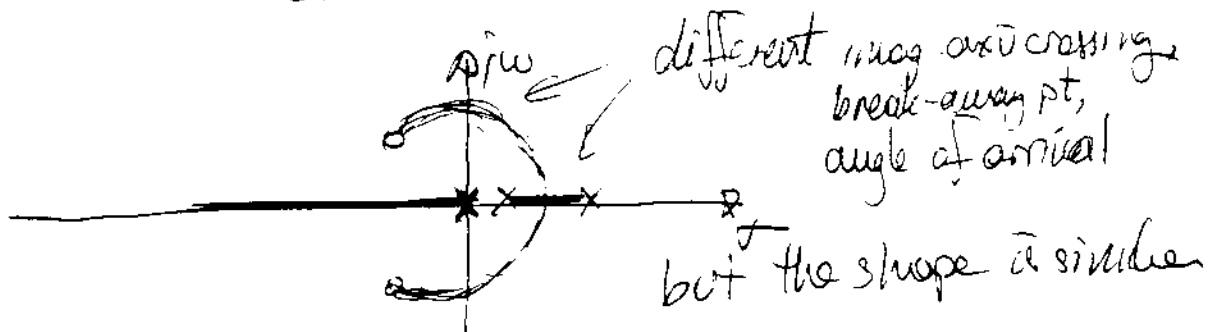
To find K from  $s_d$ , use the magnitude condition

$$\text{or } |GH| = 1, \quad \left| K \frac{s^2 + 4s + 8}{(s-3)(s-1)} \right| = 1 \Rightarrow K = 3.05$$

$s = s_d$   $s = -1 + j2.3$

by Integral Control  $\Rightarrow D(s) = \frac{K}{s}$

$$GH = K \frac{s^2 + 4s + 8}{s(s-3)(s+1)}$$



$e_{ss} = \frac{1}{K_v}$  where  $K_v = \lim_{s \rightarrow 0} s GH$  for unit ramp

$$K_v = \lim_{s \rightarrow 0} s K \frac{s^2 + 4s + 8}{s(s-3)(s+1)} = K \frac{8}{(-3)(-1)} = \frac{8}{3} K$$

$$e_{ss} = \frac{3}{8K}$$

$$e_{ss} \leq 0.1 \Rightarrow \frac{3}{8K} \leq 0.1 \Rightarrow K \geq 3.75$$

Need to check stability  $\rightarrow$  use Routh-Hurwitz criterion

$$1 + GH = 0$$

$$1 + K \frac{s^2 + 4s + 8}{s(s-3)(s+1)} = 0$$

$$s^3 + (K-4)s^2 + (4K+3)s + 8K = 0$$

$$\begin{array}{l|ll} s^3 & 1 & 4K+3 \\ s^2 & K-4 & 8K \\ s & (4K+3) - \frac{8K}{K-4} & \\ 1 & 8K & \end{array}$$

Stability  $\Rightarrow K-4 > 0, K > 4$

$$4K+3 - \frac{8K}{K-4} > 0, (K-4)(4K+3) - 8K > 0$$

$$4K^2 - 21K - 12 > 0$$

$$\hookrightarrow K = -0.52$$

$$K = 5.77$$

$$\begin{array}{c} \text{eqn} > 0, \text{eqn} < 0, \text{eqn} > 0 \\ \hline -0.52 \quad 5.77 \end{array}$$

$$\text{so } K < -0.52$$

$$\text{or } K > 5.77$$

$$8K > 0, K > 0$$

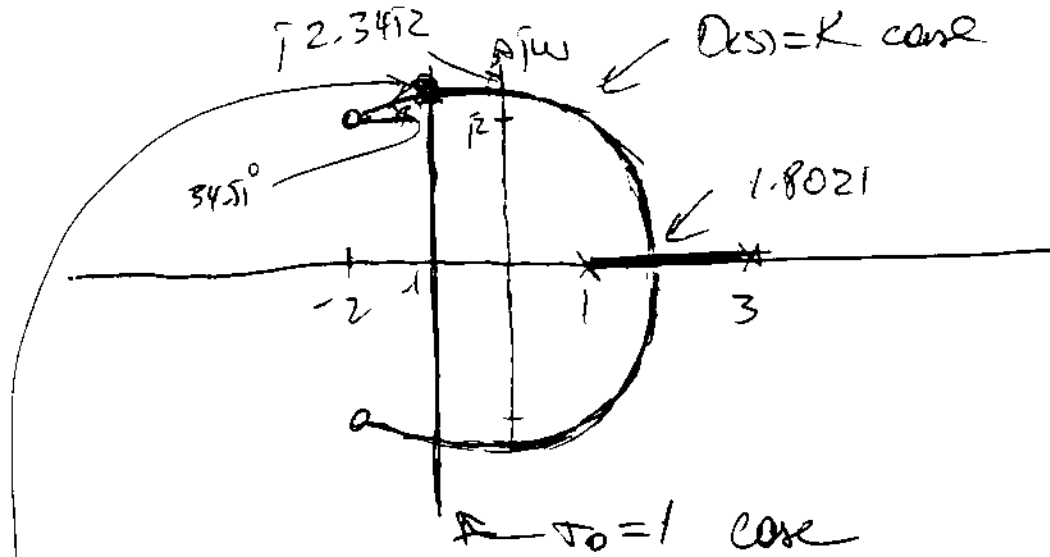
$$\Rightarrow \boxed{K > 5.77}$$

$$\text{So let } K = 6 \quad \text{or } D(s) = \frac{6}{s}$$

$$c) t_{2\%s} \leq 4 \Rightarrow \frac{4}{\sigma_0} \leq 4, \sigma_0 \geq 1$$

for a  
2nd order  
system

$$G_H = D(s) \frac{s^2 + 4s + 8}{(s-3)(s-1)}$$



Desired location  $s_d \approx -1 + j2.4$  approx.  $\rightarrow$   $M_p = e^{-\frac{\sigma_0 \pi}{\omega_d}} = e^{-\frac{1}{2.4} \pi} \approx 0.27$  or 27%

K from the magnitude condition

$$|G_H|_{s=s_d} = 1$$

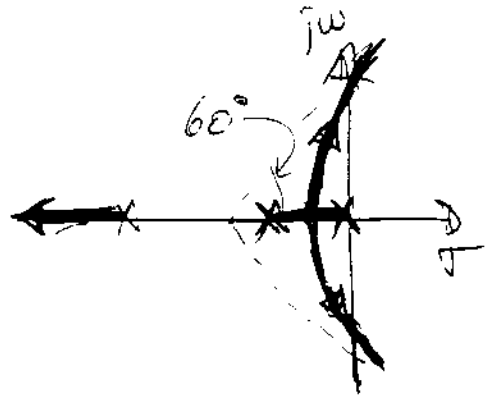
$$\left| K \frac{s^2 + 4s + 8}{(s-3)(s-1)} \right|_{s=-1+j2.4} = 1 \Rightarrow K = 2.99$$

i.e.  $t_{2\%s} \leq 4 \Rightarrow K > 2.99$ , let  $K = 3.5$



#2

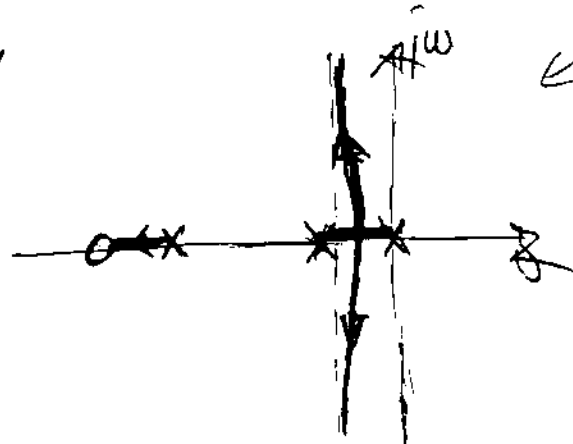
a)



← # of poles = 3  
 # of zeros = 0

$$\sigma_a = \frac{180^\circ + k360^\circ}{3-0} = \pm 60^\circ, 180^\circ$$

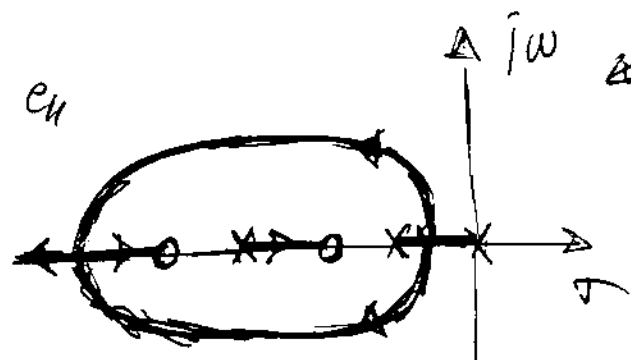
b)



← # of poles = 3  
 # of zeros = 1

$$\sigma_a = \frac{180^\circ + k360^\circ}{3-1} = \pm 90^\circ$$

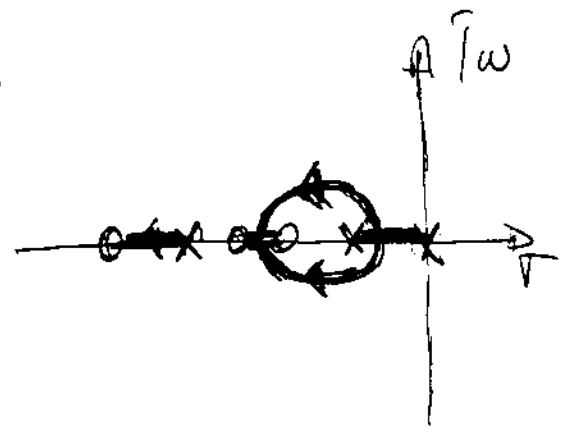
c)



← # of poles = 3  
 # of zeros = 2

$$\sigma_a = \frac{180^\circ + k360^\circ}{3-2} = 180^\circ$$

d)



← # of poles = 3  
 # of zeros = 3

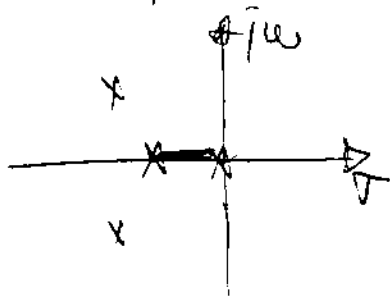
No asymptotes

McGraw-Hill  
 42 Mc  
 43 Mc  
 44 Mc  
 45 Mc  
 46 Mc  
 47 Mc  
 48 Mc  
 49 Mc  
 50 Mc  
 51 Mc  
 52 Mc  
 53 Mc  
 54 Mc  
 55 Mc  
 56 Mc  
 57 Mc  
 58 Mc  
 59 Mc  
 60 Mc

#3

$$Q(s) = K \frac{1}{s(s+2)(s^2+6s+13)} = K \frac{1}{s^4 + 8s^3 + 25s^2 + 26s}$$

a. Root locus plot  $\begin{matrix} \text{A} \\ s = -3 \pm j2 \end{matrix}$



Need to determine

- Breakaway pt
- Asymptotes
- Imag. axis crossing
- Angle of departure

→ Breakaway pt  $\rightarrow \frac{dK}{ds} = 0$

$$1 + GH = 0$$

$$1 + K \frac{1}{s^4 + 8s^3 + 25s^2 + 26s} = 0$$

$$-K = s^4 + 8s^3 + 25s^2 + 26s$$

$$-\frac{dK}{ds} = 4s^3 + 24s^2 + 50s + 26$$

$$\frac{dK}{ds} = 0 \Rightarrow 4s^3 + 24s^2 + 50s + 26 = 0$$

Solu.  $s = -0.7652$

$\text{A}$   $s = -2.6174 \pm j1.2820$

Not required

→ Asymptotes →  $\sigma_a = \frac{\sum p_i - \sum z_i}{n-m}$

$$\theta_a = \frac{180^\circ + k360^\circ}{n-m}$$

$$\sigma_a = \frac{(0 + (-2)) + (-3 + j2) + (-3 - j2)}{4-0} = -2$$

$$\theta_a = \frac{180^\circ + k360^\circ}{4-0} = \pm 45^\circ, \pm 135^\circ$$

→ Imag. axis crossing → Routh-Hurwitz criterion

$$1 + GH = 0$$

$$1 + K \frac{1}{s^4 + 8s^3 + 25s^2 + 26s} = 0$$

$$s^4 + 8s^3 + 25s^2 + 26s + K = 0$$

$$s^4 \quad | \quad 1 \quad 25 \quad K$$

$$s^3 \quad | \quad 8 \quad 26$$

$$s^2 \quad | \quad 25 - \frac{26}{8} = 21.75 \quad K$$

$$s \quad | \quad 26 - \frac{8K}{21.75}$$

$$1 \quad | \quad K$$

$$\xrightarrow{\quad} 21.75s^2 + K = 0$$

$$\rightarrow 26 - \frac{8K}{21.75} = 0 \Rightarrow K = 70.69$$

$$21.75s^2 + 70.69 = 0$$

$$s = \pm j1.8$$

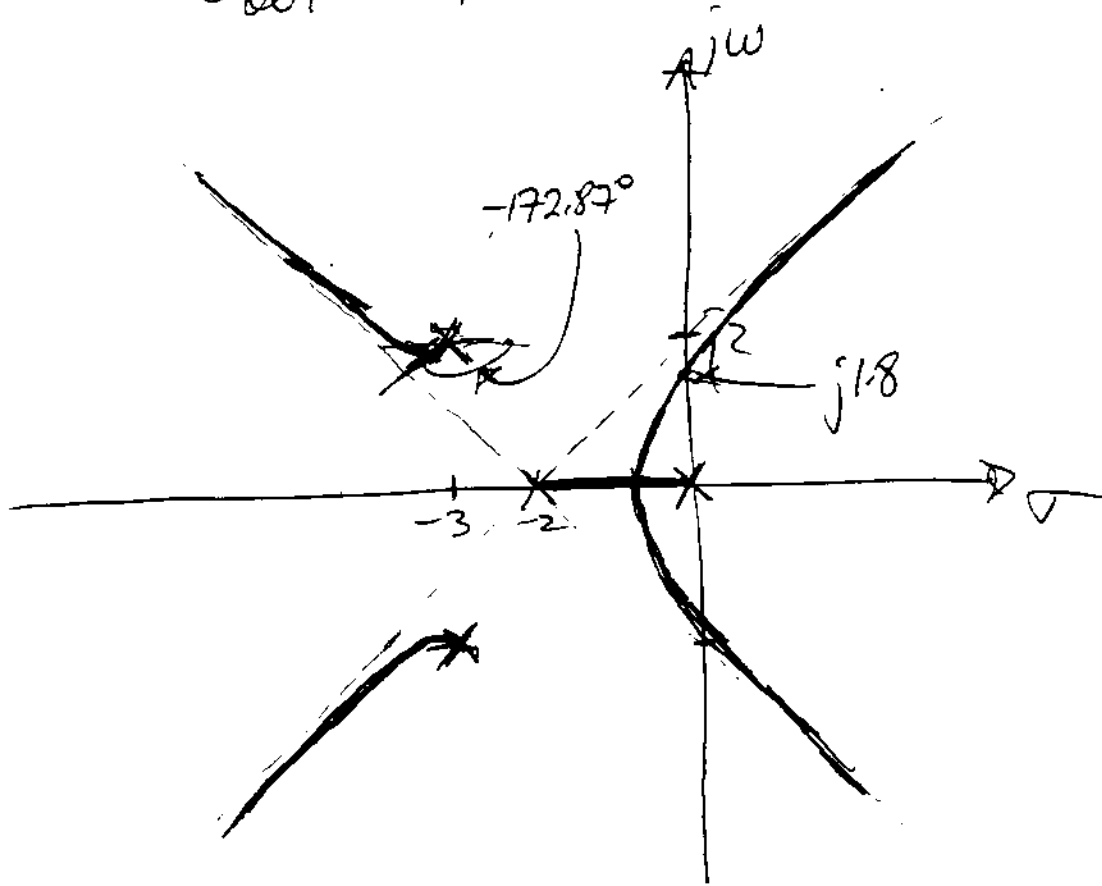
→ Angle of departure → Ang. cond about  $s_p = -3 + j2$

$$-\angle s_p - \angle (s_p + 2) - \angle (s_p + 3 + j2) - \theta_{out} = 180^\circ + k360^\circ$$

$$-\tan^{-1} \frac{2-0}{-3-0} - \tan^{-1} \frac{2-0}{-3-(-2)} - 90^\circ - \theta_{out} = 180^\circ + k360^\circ$$

$$-146.31^\circ - 116.57^\circ - 90^\circ - \theta_{out} = 180^\circ + k360^\circ$$

$$\theta_{out} = -172.87^\circ$$



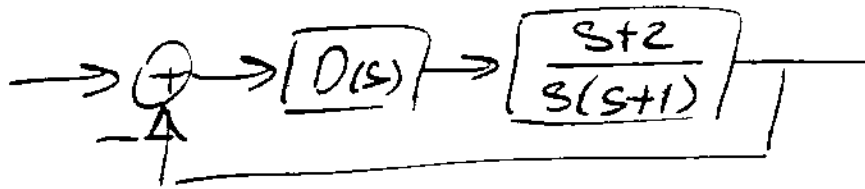
b) For stability Routh-Hurwitz table should have positive first column

$$s \rightarrow 26 - \frac{8K}{21.75} > 0 \Rightarrow K < 70.69$$

$$1 \rightarrow K \Rightarrow K > 0$$

$$\text{so } 0 < K < 70.69$$

#4

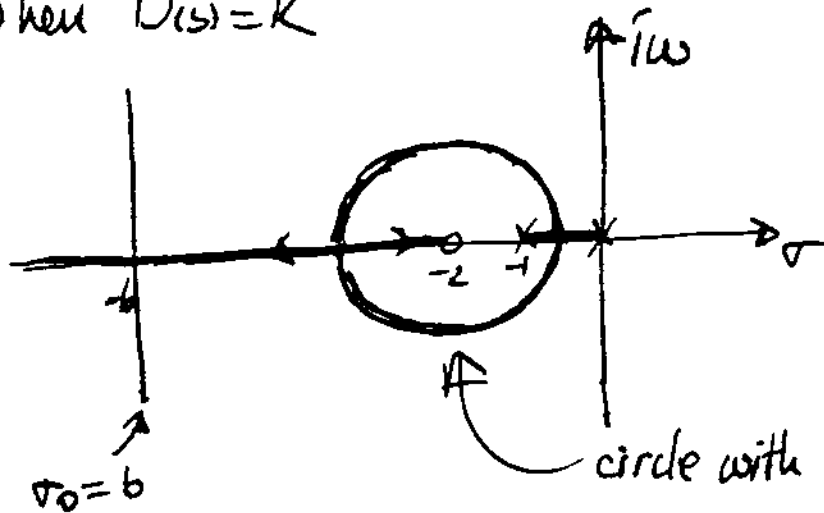


$e_{ss} = 0$  for step input  $\Rightarrow G_H = \frac{1}{s} G_1(s)$

we have already have this term in  $G_H$ , so there is no need for  $D(s)$  to supply it.

$t_{5\%s} \approx 0.5 \Rightarrow \frac{3}{\sigma_0} \approx 0.5 \Rightarrow \sigma_0 = 6$   
for a 2nd order system

When  $D(s) = K$

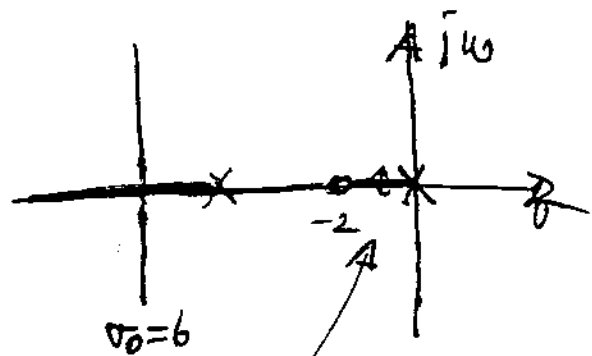


$$\text{radius} = \sqrt{(-2-0)(-2-(-1))}$$

$$= 1.4142$$

Since for  $t_{5\%s} \approx 0.5$ , one pole should have  $\sigma_0 \approx 6$  and the other one  $\sigma_0 \geq 6$ . This condition cannot be satisfied from the above root-locus diagram, since one pole approaches to  $-2$ .

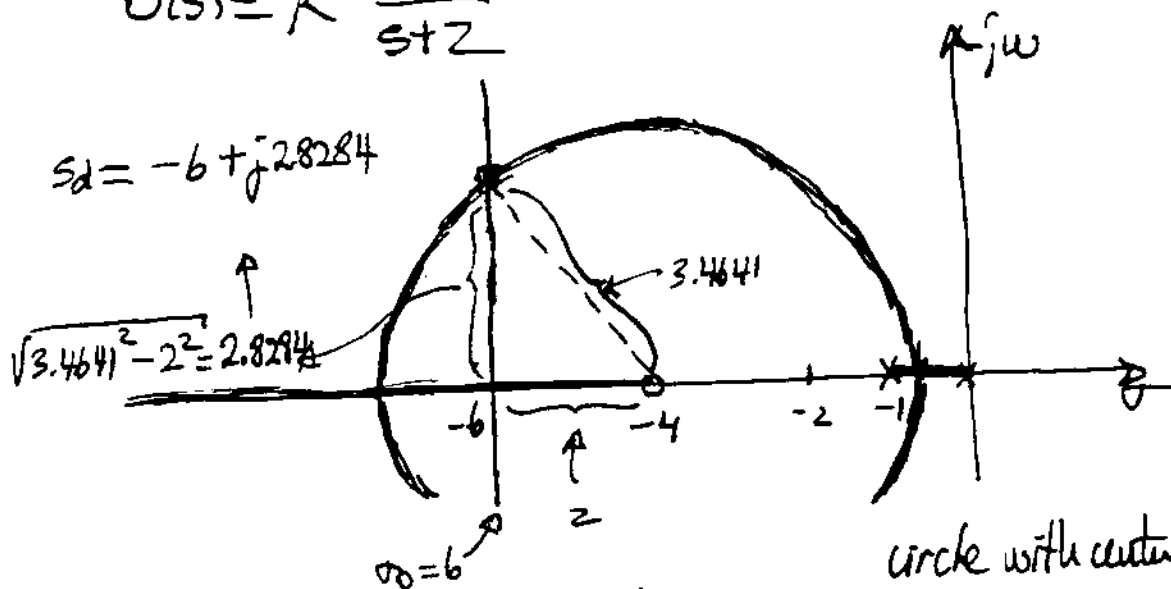
Canceled one pole and placing it any stable location does not help since the root locus becomes



this pole will always be too slow

So cancel the zero, and place it such that the circle intersect with the  $\sigma_0 = -6$ . One such choice is a zero at  $-4$ , or

$$D(s) = K \frac{s+4}{s+2}$$



circle with center =  $-4$

$$\begin{aligned} \text{radius} &= \sqrt{(-4-0)(-4+2)} \\ &= 3.4641 \end{aligned}$$

$K$  from the magnitude condition

$$|GH| = 1 \Rightarrow \left| K \frac{s+4}{s(s+2)} \right| = 1$$

$$s = -6 + j2.8284$$

$$\Rightarrow K = 9.38 \quad \text{or} \quad D(s) = 9.38 \frac{s+4}{s+2}$$